

## **Appendix B**

### **Rule 132 Declaration B - Regarding Professional Recognition of The Invention's Inventive Concepts**



**In the United States Patent and Trademark Office**

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Examiner: Berhane, Adolf D., Art Unit 2838

**Rule 132 Declaration B - Regarding Professional Recognition of  
The Invention's Inventive Concepts**

James Arthur declares as follows:

1. I am the inventor in the above patent application.
2. EDN Magazine is one of the two leading trade journals of the electronics industry for design engineers, with an international circulation.
3. EDN Magazine includes a "Design Ideas" feature wherein novel and innovative circuit designs are published.
4. In the June 12, 2003 issue of EDN Magazine, the "Design Ideas" section, pg. 88, included an article entitled "LED driver delivers constant luminosity," by Isreal Schleicher. (attached copy: Exhibit B-1)
5. The Schleicher circuit employs the inventive concepts described in Applicant's specification, substantially identically as taught therein:
  - a) an on-timer producing an on-time responsive to supply voltage  $V_{cc}$ , and
  - b) an off-time established by the discharge of the switching inductor,thereby producing a stabilized, single-cell boost converter for driving an LED.

6. I further declare that all statements made herein of my own knowledge are true and that all statements made upon information and belief are believed to be true, and further that these statements were made with the knowledge that willful false statements and the like so made are punishable by fine or imprisonment, or both, under Section 1001 of Title 18 of the United States Code, and that such willful false statements may jeopardize the validity of the application and any patent issuing therefrom.

Signature

Date

  
James Arthur

November 8, 2005

## design ideas

sure it does not saturate at the highest value of peak current. Switching transistor  $Q_1$  should have very low saturation voltage to minimize losses and produce the highest possible peak current. The addition of  $D_3$  and  $C_4$  enables the circuit to generate an auxiliary supply voltage,  $V_{AUX}$ , which you can use to drive low-power circuitry without adversely affecting the LED's intensity. With a battery voltage of 1V, the test circuit produces good light intensity in the white LED and delivers almost 1.5 mA at 4.7V to the auxiliary load. Even at  $V_{BATT}=500$  mV, the circuit delivers 340  $\mu$ A into a

10-k $\Omega$  load and maintains reasonable LED brightness. Note that  $IC_1$  cannot take power from the auxiliary rail, because  $V_{AUX}$  can easily exceed the maximum voltage rating of the two suggested device types.

The minimum start-up voltage depends largely on the device you use for  $D_1$ . Tests using a high-quality Schottky diode produce a minimum power-up voltage of just 800 mV. You can further reduce this level by replacing  $D_1$  with pnp transistor  $Q_3$  (Figure 1b). This modification allows the test circuit to start up at just 650 mV at room temperature. Note,

however, that  $Q_2$ 's collector-base junction becomes forward-biased under quiescent conditions, which results in wasted power in its base-bias resistor. Despite its simplicity, the circuit can produce spectacular results with high-brightness LEDs. The Luxeon range of LEDs from Lumileds ([www.lumileds.com](http://www.lumileds.com)) allows the circuit to demonstrate its prowess. With  $L_1$  reduced to 10  $\mu$ H and  $V_{BATT}=1$ V, the circuit generates a peak current of 220 mA in a Luxeon LXHL-PW01 white LED, resulting in dazzling light intensity. □

## LED driver delivers constant luminosity

Israel Schleicher, Bakersfield, CA

THE CIRCUIT IN Figure 1 is similar in principle to that of a previous Design Idea (Reference 1) but offers improved, more reproducible performance. The output current is almost constant over an input-voltage range of 1.2 to 1.5V and is insensitive to variations of transistor gain. Transistors  $Q_1$  and  $Q_2$  form an astable flip-flop.  $R_1$  and  $C$  define the on-time of  $Q_2$ . During that time,  $Q_1$  is off, and the voltage at the base of  $Q_1$  and the current in inductor  $L$  ramp up. When the voltage at the base of  $Q_1$  reaches approximately 0.6V,  $Q_1$  turns on, and  $Q_2$  turns off. This switching causes "flyback" action in inductor  $L$ . The voltage across the inductor reverses, and the energy stored in the inductor transfers to the LED in the form of a down-ramping pulse of current. During flyback time, voltage across the LED is approximately constant.

The voltage for yellow and white LEDs is approximately 1.9 and 3.5V, respec-

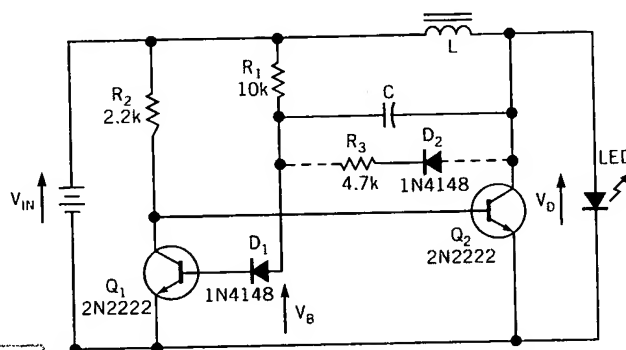


Figure 1

NOTE:  
WHITE LED REQUIRES  $R_3$  AND  $D_2$ .

This circuit delivers virtually constant luminosity for a white or a yellow LED.

tively. When the current through the LED falls to zero, the voltage at the collector of  $Q_2$  falls sharply, and this circuit condition triggers the next cycle. Assuming the justifiable approximation that the saturation voltage of  $Q_2$  is close to 0V and that the LED's forward voltage,  $V_D$ , is constant, you can easily derive the expression for the average dc current through the LED:

$$I_{AVE} = \frac{V_{IN}^2 R_1 C}{2 V_D L} \log_e \left( \frac{V_{IN} + V_D - V_B}{V_{IN} - V_B} \right)$$

At first glance,  $I_{AVE}$  depends strongly on  $V_{IN}$ . But close examination of the logarithmic term reveals that, with a proper selection of  $V_B$ , the logarithmic term can become a sharply declining function of  $V_{IN}$ . The logarithmic term thus fully compensates for the term  $V_{IN}^2$  in the expression. That compensation is precisely the purpose of the diode,  $D_1$ , in series with the base of  $Q_1$ . The circuit drives a high-brightness yellow or white LED. Table 1 shows the proper component selection for both colors. Table 1 also shows some measured results at  $V_{IN}=1.35$ V. Because the voltage across the white LED falls from 3.9 to 3.1V during flyback, capacitor  $C$  subtracts current from the amount available to the base of  $Q_1$ . This subtraction might retrigger the circuit before the current in  $L$  falls to zero. The addition of  $R_3$  and  $D_2$  solves this problem. During flyback, the current that flows through  $R_3$  compensates for the current withdrawn through  $C$ . □

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### COMPONENT SELECTION FOR YELLOW OR WHITE LED

LED	L (mH)	C (pF)	$D_1$	Current drain (mA)	LED current (mA)	Frequency (kHz)	Power-conversion efficiency (%)
Yellow	1	470	1N4003	5.6	$3.3 \pm 0.1$	40	83
White	2	1800	1N752	12.4	$3.7 \pm 0.2$	15	78

### REFERENCE

1. Nell, Susanne, "Voltage-to-current converter drives white LEDs," *EDN*, June 27, 2002, pg 84.